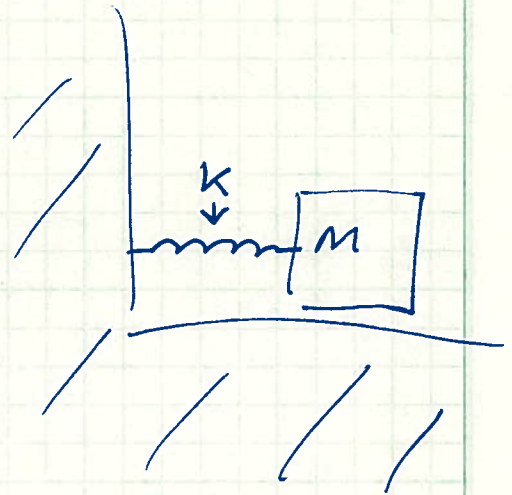
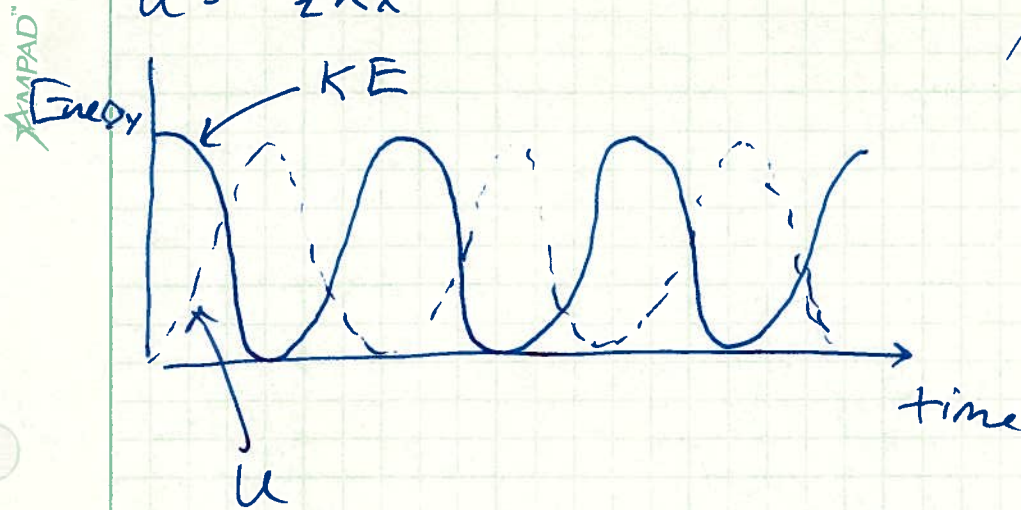


Electrical Oscillators

Mechanical Oscillator: energy is converted between kinetic and elastic potential

$$KE = \frac{1}{2}mv^2 = \frac{1}{2}m\dot{x}^2$$

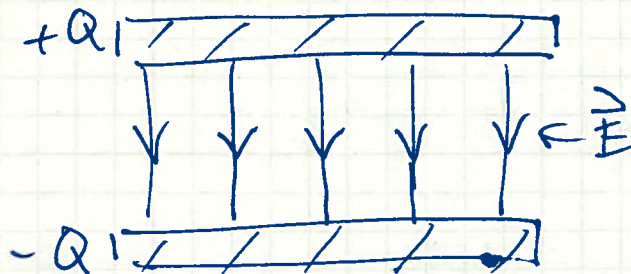
$$U = \frac{1}{2}kx^2$$



Electrical Oscillator: Energy is converted between electric ( $\vec{E}$  field) and magnetic ( $\vec{B}$  field).

① Capacitor: Device for storing energy in an electric field

Ex: Parallel Plate Capacitor



Each small volume of space ( $dV$ ) with an electric field  $\vec{E}$  stores a small amount of electric energy ( $dU_E$ ):

$$dU_E = \underbrace{\frac{1}{2} \epsilon_0 |\vec{E}|^2}_{u_E} dV, \quad u_E = \frac{1}{2} \epsilon_0 |\vec{E}|^2$$

Energy, big  $U$   $\nearrow$

$u_E$   
↑  
energy density, little  $u$

= "electric energy density of free space"  
=  $\frac{\text{Joules}}{\text{meter}^3}$

The total energy stored is

"big  $U$ "  $\rightarrow U_E = \int_{\text{all space}} \frac{1}{2} \epsilon_0 |\vec{E}|^2 dV$

If the electric field is created by a capacitor with charges  $+Q$  and  $-Q$  and voltage difference  $V$ , then the total energy can be written

$$U_E = \frac{1}{2} QV = \frac{1}{2} CV^2 = \frac{Q^2}{2C}$$

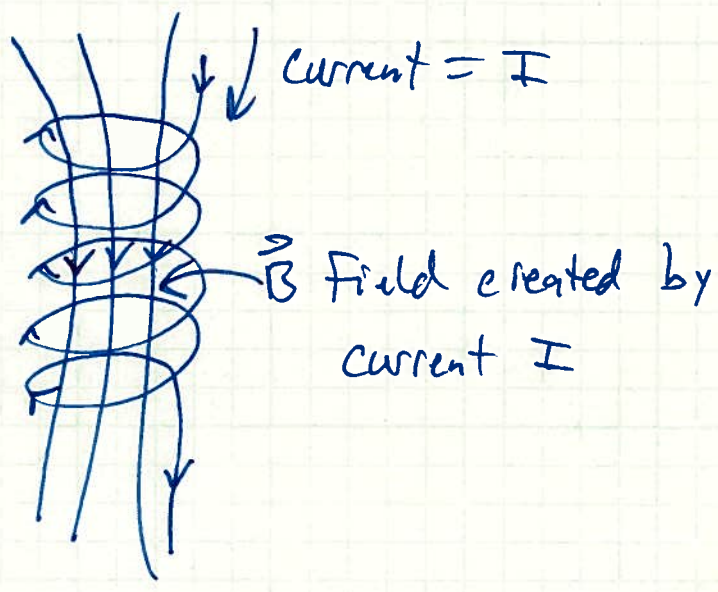
where  $C = \text{capacitance} = \frac{Q}{V}$

Capacitance is a constant that only depends on the shape and material of the capacitor.

" $Q = CV$ " says "charge on the capacitor is proportional to the voltage across it.  $C$  is the proportionality constant."

② Inductor: Device for storing energy in a magnetic field.

A simple solenoid:



The energy density of a magnetic field is

$$dU_B = \frac{1}{2\mu_0} |\vec{B}|^2 dV, \quad u_B = \frac{1}{2\mu_0} |\vec{B}|^2$$

The total magnetic energy is

$$\Rightarrow U_B = \int_{\text{all space}} \frac{1}{2\mu_0} |\vec{B}|^2 dV$$

"little u" = "magnetic energy density of free space"

"big u"

For an inductor, the total energy can be written as

$$= \frac{\text{Joules}}{\text{meter}^3}$$

$$U_B = \frac{1}{2} LI^2, \text{ where } L = \text{"self-inductance"}$$

L is determined by the shape and material of the inductor. It is the proportionality constant between magnetic flux and current:

$\Phi_B$  = magnetic flux through the inductor =  $LI$   
 ↑ ↓ current  
 proportionality constant.

Similarities between C & L

Device	Circuit Symbol	MKS unit	Stores energy in:	Proportionality Constant:	Determined by
C		Farad	$\vec{E}$	$Q = CV$	Shape and material
L		Henry	$\vec{B}$	$\Phi_B = LI$	

If you know the shape & material of your capacitor/inductor, then you can calculate  $\left\{ \begin{matrix} C \\ L \end{matrix} \right\}$ .

Neither C nor L depends on  $\left\{ \begin{matrix} Q \\ V \\ I \end{matrix} \right\}$ . These things depend on time, but C & L are constant.

Voltage Rules:

Capacitor:  $|V_C| = \left| \frac{1}{C} Q \right|$  or  $V_C = \frac{1}{C} Q$  (ignoring any sign)

Inductor:  $|V_L| = \left| - \frac{d\Phi_B}{dt} \right| = \left| - \frac{d(LI)}{dt} \right| = \left| L \frac{dI}{dt} \right|$

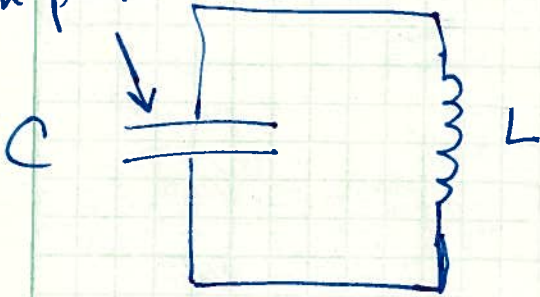
$\therefore V_L = L \frac{dI}{dt}$  (ignoring any sign)

LC oscillator

Simplest electrical oscillator.

Energy exchanges between electric & magnetic.

q = charge on plate



Voltage Loop rule:

$$V_C + V_L = \emptyset$$

$$\left(\frac{1}{C} q\right) \left(L \frac{dI}{dt}\right) = L \frac{dq}{dt^2}$$

since  $I = \frac{dq}{dt}$

$$\therefore \frac{1}{C} q + L \frac{d^2 q}{dt^2} = \emptyset$$

$$\ddot{q} + \frac{1}{LC} q = \emptyset$$

Simple harmonic oscillator equation

Solution:  $q(t) = q_0 e^{i(\omega_0 t + \delta)}$ , where  $\omega_0 = \frac{1}{\sqrt{LC}}$

$q_0$  &  $\delta$  are determined by the initial conditions.

= "natural freq."

$$\dot{q}(t) = I(t) = (i\omega_0)(q_0 e^{i(\omega_0 t + \delta)}) = i\omega_0 q(t)$$

↑  
current has a phase shift of 90° compared to charge.

Energy:  $U_E = \text{electric energy}$

$$= \frac{1}{2C} q^2 = \frac{1}{2C} [q_0 \cos(\omega_0 t + \delta)]^2$$

$$= \frac{q_0^2}{2C} \cos^2(\omega_0 t + \delta)$$

$U_B = \text{magnetic energy}$

$$= \frac{1}{2} L I^2$$

$$= \frac{1}{2} L \left[ \frac{q_0}{C} \operatorname{Re} \left( i \omega_0 q_0 e^{i(\omega_0 t + \delta)} \right) \right]^2$$

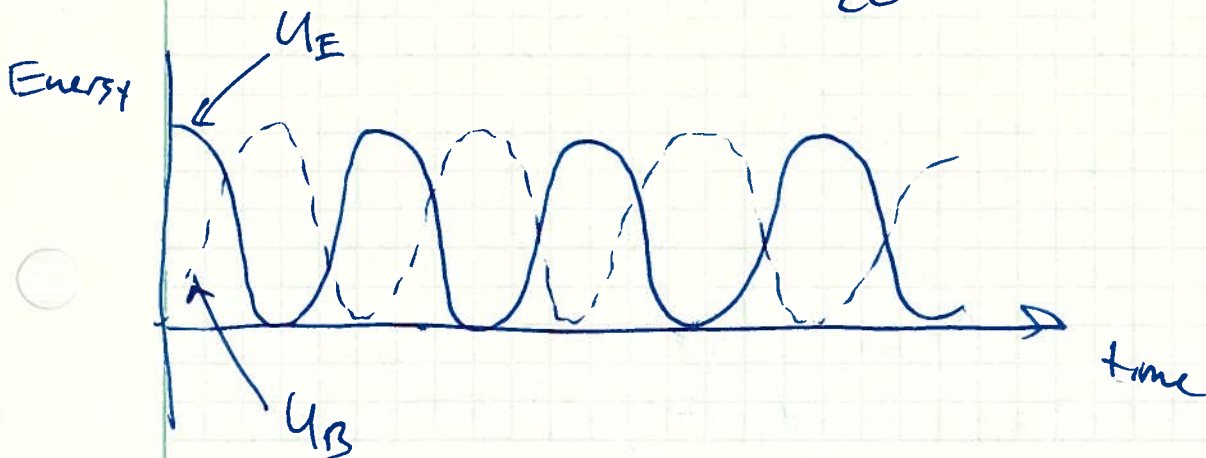
$$= \frac{1}{2} q_0^2 \omega_0^2 L \sin^2(\omega_0 t + \delta)$$

$$\omega_0^2 = \frac{1}{LC} \quad \text{so} \quad \omega_0^2 L = \frac{1}{C}$$

$$= \frac{q_0^2}{2C} \sin^2(\omega_0 t + \delta)$$

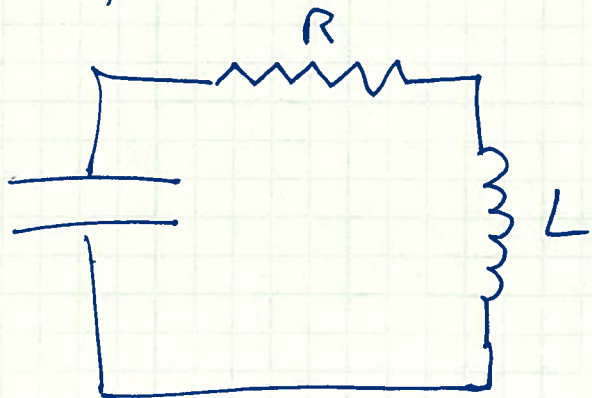
Total energy =  $U_E + U_B = \frac{q_0^2}{2C} \left[ \cos^2(\omega_0 t + \delta) + \sin^2(\omega_0 t + \delta) \right]$

$$= \frac{q_0^2}{2C} = \text{constant}$$



# LC oscillator with damping - RLC circuit

Add a resistor to the circuit: Electrical energy will be converted to heat in the resistor



Voltage Rule:

$$V_C + V_R + V_L = 0$$

$$\downarrow \quad \downarrow \quad \downarrow$$

$$\frac{q}{C} \quad IR \quad L \frac{dI}{dt} = L \frac{dq}{dt}$$

$$= -\dot{q}R \quad = L\ddot{q}$$

$$\therefore \frac{q}{C} + R\dot{q} + L\ddot{q} = 0$$

$$\ddot{q} + \frac{R}{L}\dot{q} + \frac{1}{LC}q = 0$$

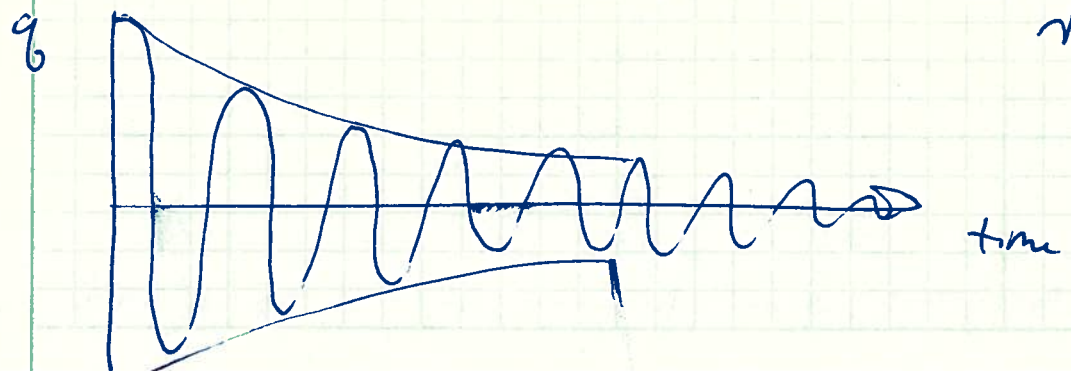
Simple Harmonic Oscillator with damping

Solution (light damping):

$$q(t) = q_0 e^{-\gamma/2 t} e^{i(\omega t + \delta)}$$

where  $\gamma = \frac{R}{L}$

and  $\omega = \sqrt{\omega_0^2 - \frac{\gamma^2}{4}}$



# Driven RLC circuit - Series

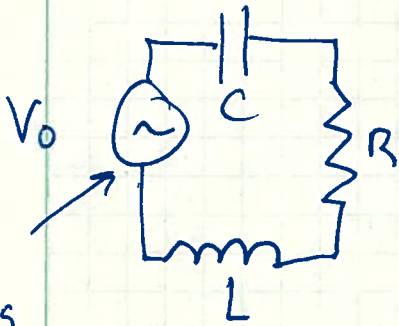
Suppose we have an oscillating circuit with a voltage source that varies in time as a cosine:

$$V_{\text{source}}(t) = V_0 \cos(\omega t) = V_0 e^{i\omega t}$$

↑  
driving frequency

We choose  $\delta$  (phase shift) =  $\phi$  by choosing  $t = \phi$  correctly.

Then the RLC series circuit looks like:



time varying voltage source

Voltage Loop Rule:

$$V_S = V_C + V_R + V_L$$

$\downarrow$                      $\downarrow$                      $\downarrow$                      $\downarrow$   
 $(V_0 e^{i\omega t})$      $(\frac{1}{C} q)$      $(R \dot{q})$      $(L \ddot{q})$

$$\ddot{q} + \left(\frac{R}{L}\right) \dot{q} + \left(\frac{1}{LC}\right) q = \left(\frac{V_0}{L}\right) e^{i\omega t}$$

Driven Harmonic Oscillator.

Steady State

Solution:  $q(t) = q_0 e^{i(\omega t + \delta)}$  or  $A e^{i(\omega t + \delta)}$

where

$$q_0(\omega) = \frac{V_0/L}{\sqrt{(\omega_0^2 - \omega^2)^2 + (\omega r)^2}}$$

$\omega$  = driving frequency

$$\omega_0 = \frac{1}{\sqrt{LC}}$$

$$r = (R/L)$$

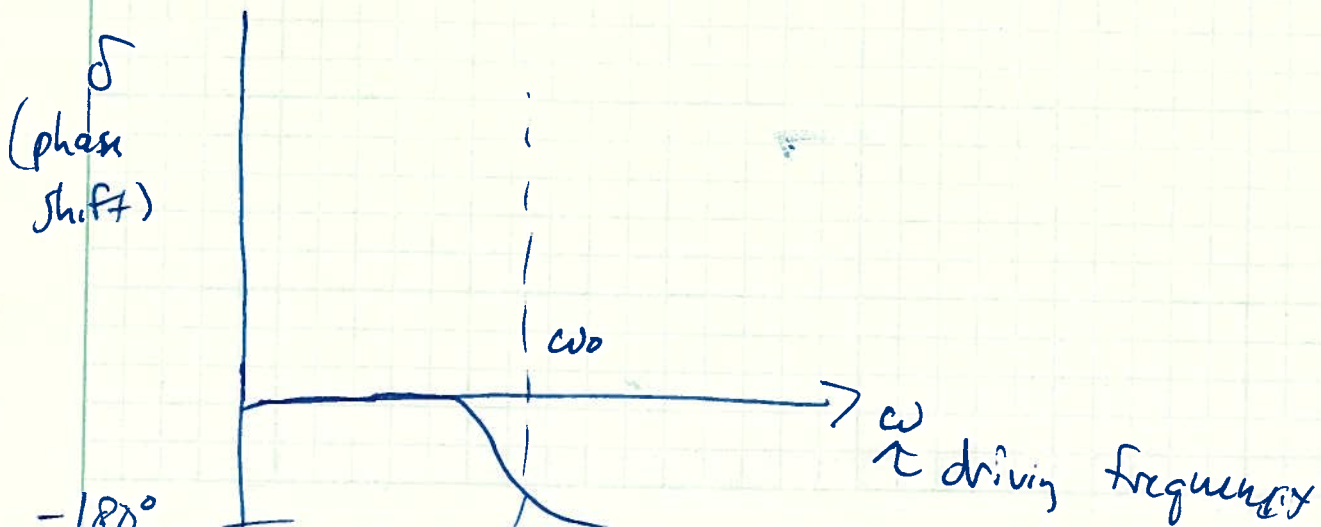
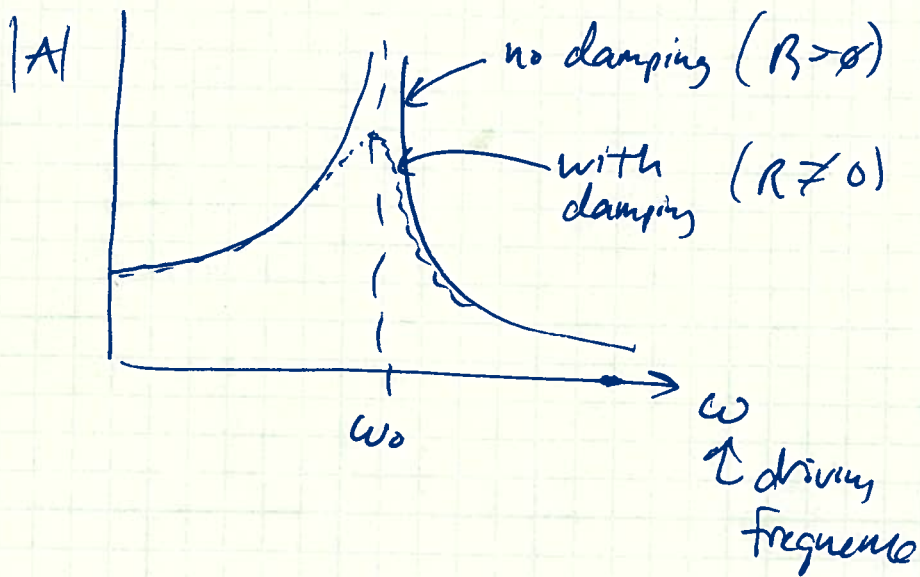


and  $\delta(\omega) = -\tan^{-1} \left[ \frac{\omega R}{\omega_0^2 - \omega^2} \right]$

Just like the forced mechanical oscillators, we have a resonance when  $\omega \approx \omega_0$ .

When  $\omega \approx \omega_0$ , the following things become very large:

- 1) the peak charge on the capacitor
- 2) the peak current in the circuit
- 3) the <sup>total</sup> energy stored in the C & L.



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## Phasor Analysis of AC circuits

Phasors give us a geometric method for thinking about harmonic oscillators. Particularly useful for AC circuits.

### Basic Principles

① Resistors: Voltage Rule:  $V = IR$

If  $V_R = V_0 e^{i\omega t}$ , then  $I = \frac{V_0}{R} e^{i\omega t}$

$V_R = IR$  Ohm's Law for resistors

no phase shift

② Capacitors:  $Q = CV$

$$I = \frac{dQ}{dt} = C \frac{dV}{dt}$$

If  $V = V_0 e^{i\omega t}$ , then  $I = i\omega C V_0 e^{i\omega t}$

or

$$V = I \left( \frac{1}{i\omega C} \right)$$

Define  $X_C \equiv \frac{1}{i\omega C}$

"Capacitive Reactance"

The

$V = IX_C$

"Ohm's Law for Capacitors"

(2)

③ Inductors:  $V = L \frac{dI}{dt}$

If  $V = V_0 e^{i\omega t}$

Then  $I = \frac{1}{i\omega L} (V_0 e^{i\omega t})$

OR

$$V = I (i\omega L)$$

Define  $i\omega L \equiv X_L =$  "inductive reactance."

Then  $V = IX_L$  Ohm's Law  
for Inductors

Summarizing

Resistors:  $V = IR$ ,  $\Rightarrow V$  &  $I$  are 100% in phase  
(no phase difference)

Capacitors:  $V = IX_C$ ,  $X_C = \frac{1}{i\omega C} = \frac{-i}{\omega C}$  ← phase difference  
 $\Rightarrow$  Voltage lags the current  
by  $90^\circ$  ( $\pi/2$ )

Inductors:  $V = IX_L$ ,  $X_L = i\omega L$  ← phase difference  
 $\Rightarrow$  Voltage leads the current  
by  $90^\circ$ .

In AC circuits, Capacitors and Inductors behave like resistors in that the peak voltage is proportional to the peak current. (Like Ohm's Law for resistors.) On the other hand,

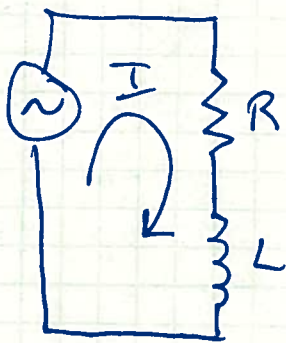
① There is a phase shift of  $+90^\circ$  or  $-90^\circ$  between voltage & current.

② The proportionality "constant" ~~for~~ depends on the frequency:  $X_C = \frac{-i}{\omega C}$ ,  $X_L = i\omega L$ .  
frequency dependent

### Application to Phasor Analysis

Ex:

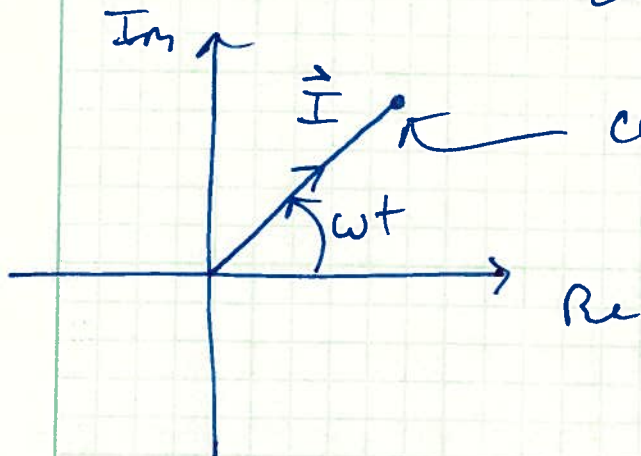
$$V_s = V_0 e^{i\omega t}$$



Driven RL circuit.

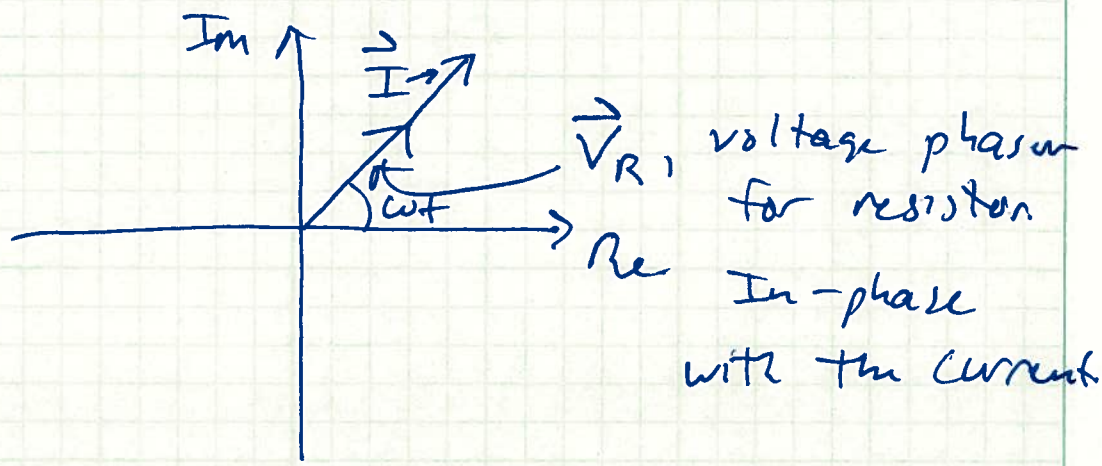
This circuit has one and only one current.

Let's draw it in the complex plane:

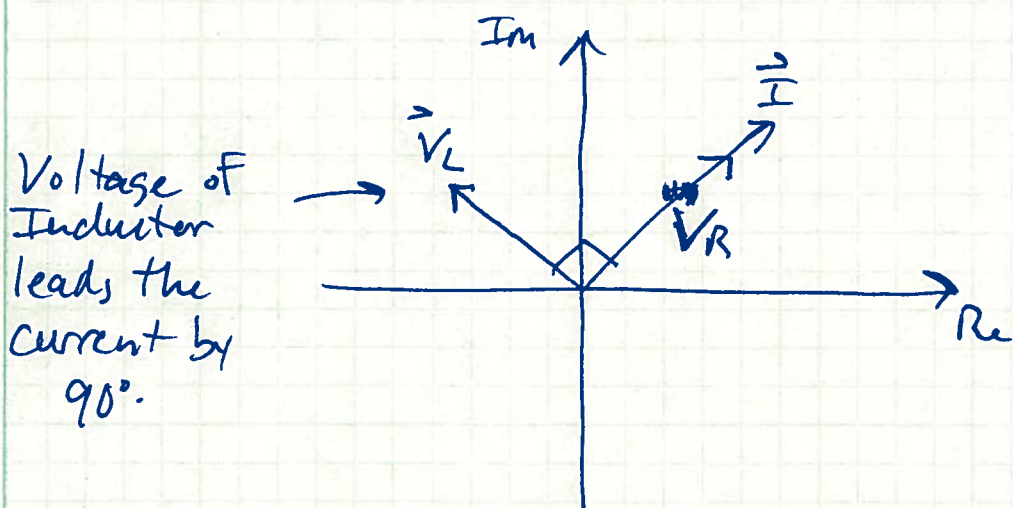


current phasor, rotating at frequency  $\omega$ . Makes an angle  $\omega t$  with the real axis.

Add the voltage phasor for the resistor



Now add the voltage phasor for the inductor:

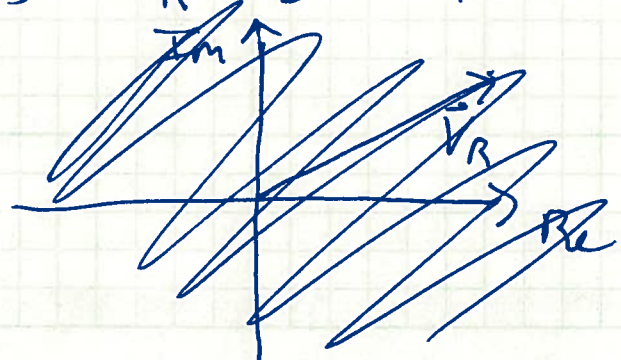


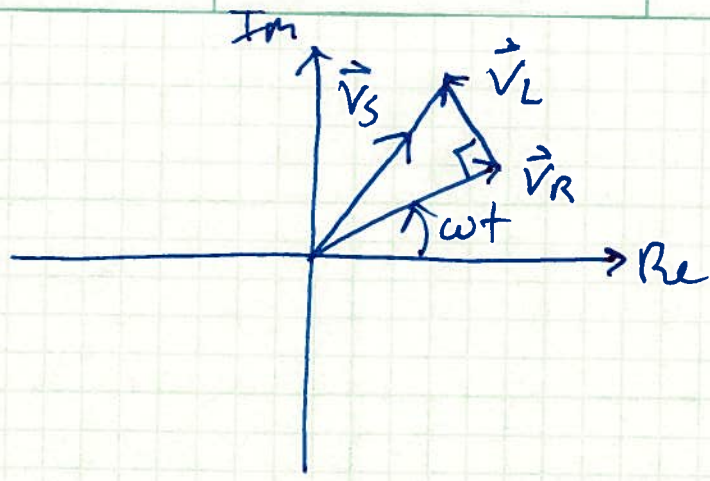
Now the voltage loop rule says:

$V_s = V_R + V_L$  which we can interpret geometrically.

$\vec{V}_s = \vec{V}_R + \vec{V}_L$  ← phasors

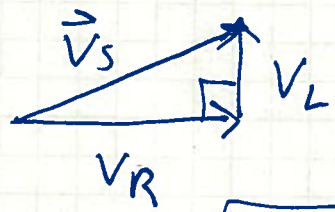
or





This is a phasor diagram for the voltages in the circuit

This is useful because we can use the geometry to figure out the relationships:



$$\therefore |\vec{V}_s| = V_0 = \sqrt{|V_R|^2 + |V_L|^2}$$

$$|V_R| = I_0 R$$

$$|V_L| = I_0 |i\omega L| = I_0 (\omega L)$$

$$V_0 = \sqrt{(I_0 R)^2 + (\omega L I_0)^2}$$

$$V_0 = I_0 \left[ \sqrt{R^2 + (\omega L)^2} \right] = I_0 Z$$

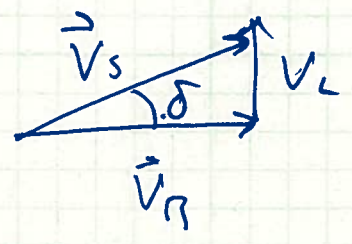
Define  $Z = \text{"impedance"}$

Peak current  $I_0 = \frac{V_0}{\sqrt{R^2 + (\omega L)^2}}$

Peak current drops as driving frequency increases



How about phase differences?



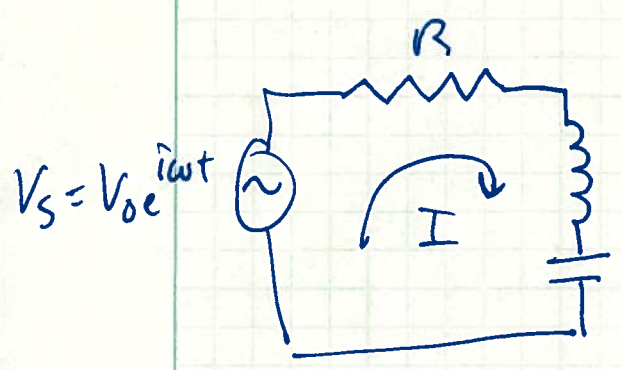
phase difference between  $V_R$  &  $V_s$ : (or  $I$  &  $V_s$ )

$$\delta = \tan^{-1} \left( \frac{|V_L|}{|V_R|} \right) = \tan^{-1} \left( \frac{\omega L I_0}{R I_0} \right)$$

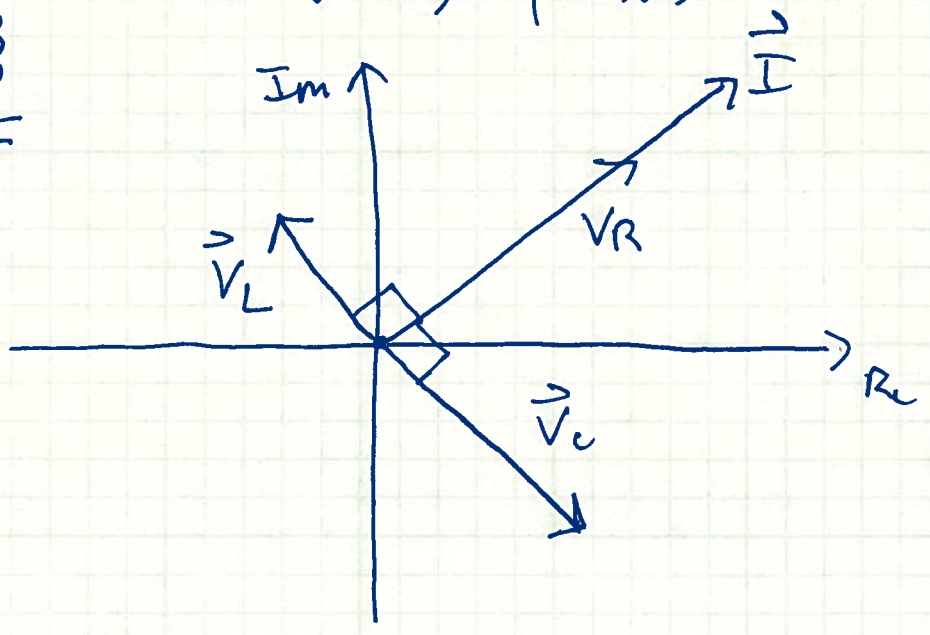
$$\delta = \tan^{-1} \left( \frac{\omega L}{R} \right)$$

Example

Driven RLC circuit:

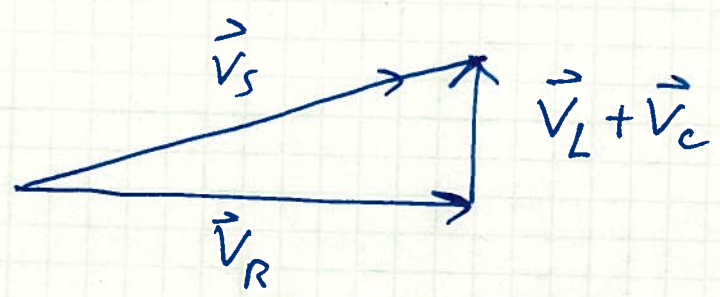


One current phasor,  
four voltage phasors:



Voltage Loop Rule:  $\vec{V}_s = \vec{V}_R + \vec{V}_L + \vec{V}_C$

Note that  $\vec{V}_L$  &  $\vec{V}_C$  are in opposite directions.



$$\begin{aligned} \therefore |\vec{V}_s| = V_0 &= \sqrt{|V_R|^2 + (\vec{V}_L + \vec{V}_C)^2} \\ &= \sqrt{(I_0 R)^2 + \left(I_0 \omega L - \frac{I_0}{\omega C}\right)^2} \\ &= I_0 \sqrt{R^2 + \left(\omega L - \frac{1}{\omega C}\right)^2} \end{aligned}$$

Z = impedance of the circuit

$$\begin{aligned} \therefore I_0 = \text{peak current} &= \frac{V_0}{\sqrt{R^2 + \left(\omega L - \frac{1}{\omega C}\right)^2}} \\ &= \frac{(V_0/L)}{\sqrt{\left(\frac{R^2}{L^2}\right) + \left(\omega - \frac{1}{\omega LC}\right)^2}} \\ &= \frac{R}{L} = \gamma \quad , \quad \frac{1}{LC} = \omega_0^2 \end{aligned}$$

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$$I_0 = \frac{\omega (V_0/L)}{\sqrt{(\omega R)^2 + (\omega^2 - \omega_0^2)^2}}$$

Resonance when  $\omega \approx \omega_0$ .

Previously we found by solving the differential equation that

$$q_0 = \frac{(V_0/L)}{\sqrt{(\omega R)^2 + (\omega^2 - \omega_0^2)^2}} \quad \text{or} \quad q(t) = q_0 e^{i\omega t}$$

which means that

$$I(t) = \dot{q}(t) = i\omega q_0 e^{i\omega t}$$

$\underbrace{\hspace{2em}}_{I_0 = \omega q_0}$

$$\text{or} \quad I_0 = \frac{\omega (V_0/L)}{\sqrt{(\omega R)^2 + (\omega^2 - \omega_0^2)^2}}$$

so we got the same result without calculating with the differential equation.

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